INSTRUCTION MANUAL

## Type 1666

Dc Resistance
Bridge

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## Warranty

We warrant that this product is free from defects in material and workmanship and, properly used, wil perform in full accordance with applicable specifications. If, within a period of ten years after original shipment, it is found, ofter examination by us or our authorized representative, not to meet this standard, it will be repaired or, at our option, replaced as follows:

* No charge for parts, labor or transportation during the first three months after original shipment:
* No charge for parts or labor during the fourth through the twelfth month after original shipment for a product returned to a GR service facility:
* No charge for parts during the second year after original shipment for a product retumed to a GR service facility:
- During the third through the tenth year after original shipment, and as long thereafter as parts are available, we will maintain our repair capability and it will be available at our then prevailing schedule of charges for a product returned to a GR service facility.

This warranty shall not apply to any product or part thereof which has been subject to accident, negligence, alteration, abuse or misuse; nor to any parts or components that have given normal service. This warranty is expressly in lieu of and excludes all other warranties expressed or implied, including the warranties of merchantability and fitness for a particular purpose; and all other obligations or liabilities on our part, Ineluding liability for consequential damages resulting from product failure or other causes, No person, firm or corporation is authorized to assume for us any other liability in connection with the sale of any product.

## Specifications

Bridge Circuits: Kelvin and guarded Wheatstone in both resistance and conductance configurations.
Ranges: TOTAL MEASUREMENT RANGE, $1 \mu \Omega$ to 1 T $\Omega$. Resistance ranges, $1 \mu \Omega$ to $1.1 \mathrm{M} \Omega$ in 7 ranges ( $1 \mu \Omega$ is one count), conductance ranges, 1 p 3 to 1.1 U in 7 ranges (1 p 3 is one count). RECOMMENDED RANGES; Wheatstone, $100 \Omega$ to $1 \mathrm{~T} \Omega ;$ Kelvin, $1 \mu \Omega$ to $10 \mathrm{k} \Omega$.
Resolution: Six digits or $1,111,110$ counts. $\quad T^{* 2 p t} 50 / 6$
Accuracy (limit of error) DIRECT READING, $\pm 0.01 \%+10 \mathrm{ppm}$ of full scale). For low-value readings, when first and second digits are zero, $\pm(0.1 \%+3 \mathrm{ppm}$ of full scale). These limits apply from 20 to $25^{\circ} \mathrm{C}$ at $<75 \% \mathrm{RH}$, within 6 months of calibration. Error remains less than $\pm 0.1 \%$ from 0 to $25^{\circ} \mathrm{C}$ at $95 \% \mathrm{RH}$ and from 0 to $35^{\circ} \mathrm{C}$ at $85 \%$ RH. LONG TERM ACCURACY: Add $0.01 \%$ per year, if not recalibrated. COMPARISON ACCURACY: $\pm[2+0.001 \times 1 \mathrm{ppm}$ difference)] ppm of full scale (decade values to 2 ppm where sensitivity is adequate and difference is small).
Sensitivity (with internal source): RESISTANCE: $2 \mu \Omega$ at very low values; 10 ppm at $1 \Omega ; 5$ ppm at $10 \Omega ; 1 \mathrm{ppm}$ at $0.1,1,10$, and $100 \mathrm{k} \Omega$; 5 ppm at $1 \mathrm{M} \Omega$. CONDUCTANCE: $2 \mathrm{p} \mho$ at very low values, 5 ppm at $1 \mu \mho ; 1 \mathrm{ppm}$ at 10 and $100 \mu \mathrm{~J}, 1$ and $10 \mathrm{mw} ; 5 \mathrm{ppm}$ at $100 \mathrm{~m} \% ; 10 \mathrm{ppm}$ at 1 v . An external source can be used for even better sensitivity.

Sources: INTERNAL: 6 V (set of 4 D cells), 0.01 W max for resistance bridge. EXTERNAL: Up to $30 \mathrm{~V} \mathrm{dc}, 0.5 \mathrm{~W}$ max.
Detector: SENSITIVITY: Mieter deflection $\approx 5 \mathrm{~mm} / \mu V$. INPUT RESISTANCE: approx $20 \mathrm{k} \Omega$. SHORT-CIRCUIT NOISE (slow position): Approx $0.1 \mu \mathrm{~V}$ pk-pk. DRIFT: Typically $0.5 \mu \mathrm{~V} / \mathrm{h}$. RESPONSE (slow/normal/fast, respectively): Low-level time constant, 4/2.5/0.7 s; high-level meter reversal, 1/0.5/0.3 s.
Guard (Wheatstone): No error with $>5 \mathrm{M} \Omega$ to ground, either terminal.
Lead Error (Kefvin): Less than $2 \mu \Omega$ additional with $\leqslant 0.1 \Omega$ in any lead.
Supplied: Set of 4 leads with goid-plated copper alligator clips. Avaitable: 1440 Standard Resistors, for recalibration.
Power: Battery of 8 D cells (Burgess type 1200 or equivalent), i.e., 4 for internal bridge source and 4 for detector power.
Wechanical: Flip-Tilt case. DIMENSIONS (wxhxd): $15 \times 12 \times 8 \mathrm{in}$. ( $381 \times 305 \times 203 \mathrm{~mm}$ ). WEIGHT: $21 \mathrm{lb}(10 \mathrm{~kg})$ net.

## Description : Catalog

$1666-9700$


Figure 1-1. Type 1665 De Resistance Bridge.

## Introduction-Section 1

1.1 DESCRIPTION ..... 1.1
1.2 OPENING AND TILTING THE CABINET ..... $1-1$
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### 1.1 DESCRIPTION.

The Type 1666 D-c Resistance Bridge (Figure 1-1) is a self-contained resistance measuring system that includes Wheatstone and Kelvin bridges for the measurement of resistance and conductance. Features of this bridge include a basic $.01 \%$ accuracy, convenient lever switches, visual null indications, complete portability, and a convenient carrying case.

### 1.2 OPENING AND TILTING THE CABINET.

The directions for opening the Type 1666 D-c Resistance Bridge are given on the panel at the rear of the instrument. Once open, the instrument can be tilted to any convenient angle. The angle should be chosen to give the most comfortable access to the controls and the best view of the meter.

Whether the instrument is open or closed, the cover forms a convenient storage place for the instruction manual
and for any other test data that should be kept with the instrument.

### 1.3 POWER SUPPLY.

The 1666 is powered by 8 D cells: 4 cells are used to power the detector, and 4 are for power to the bridge (source). Tubes containing the batteries are marked to indicate the proper placement for polarity. With the instrument removed from its cabinet, the top tube is for the source; the bottom for the detector.

To install the batteries proceed as follows:
a. Open the instrument cabinet until the instrument and cover are at $90^{\circ}$.
b. Remove the 2 cabinet screws located on the rear of the cabinet.
c. Lift the instrument from its cabinet.
d. To remove either battery tube, push the tube in the direction of the arrow (toward the spring), until the opposite end is clear of the plus ( + ) terminal, and lift the tube out.
e. Insert 4 batteries in each tube, observing the correct polarity flat (negative) ends of the batteries towards the spring and place the battery tube in its holder.
f. Mount the instrument in its cabinet and install the cabinet screws.

### 1.4 SYMBOLS, ABBREVIATIONS, DEFINITIONS.

The following symbols, abbreviations, and definitions are used in this instruction manual:
G. conductance, the inverse of resistance; $G=\frac{1}{R}=\frac{1}{E}$
$G_{x}$ unknown conductance
$R$ resistance, the ratio of voitage to current; $R=\frac{E}{1}(d c)$
$R_{A}$ decade resistance
$R_{B}$ standard resistor
$R_{r}$ ratio arm resistance
$R_{X}$ unknown resistance
$\Omega$ ohm, a unit of resistance
kilohm, $1 \mathrm{k} \Omega=1000$ ohms

$$
M \Omega
$$

megohm, $1 \mathrm{M} \Omega=1 \times 10^{6}$ ohms
$\vartheta$ mho, a unit of conductance
case
$\stackrel{1}{2}$
ground
ms
$\mu \approx$
$p \mho$ picomho, $1 p \mho=1 \times 10^{-12}$ mho

### 1.5 RESISTANCE DECADE READINGS.

The bridge balance contains 6 lever-type decades in series. Each decade has 11 digits (from 0 to 10). An $X$ is used to indicate the 10, to avoid the extra decimal place. All but the very highest of readings can be obtained without using the $X$ position (which is added mainly to facilitate the balancing procedure), so that it need not be used. However, a little practice in interpreting readings containing one or more X's will allow faster balancing, particularly with unknown values containing several zeroes. Some examples:

$$
\begin{gathered}
37123 X=371240 \\
6842 \times 4=684304 \\
761 \times X X=762110 \\
769 \times 4 \times=76 \times 050=770050 \\
X \times X \times X=1111110
\end{gathered}
$$

### 1.6 CONNECTIONS.

The bridge terminals ( $A, B, C$ and $D$ ) and the ground terminal are gold-plated copper binding posts that accept banana plugs, standard telephone tips, alligator clips, spade terminals and any wire size up to 11 AWG, as shown in Figure 1-2. Copper binding posts are used to minimize thermally induced voltages when connected to copper wire. The test lead set supplied has copper alligator clips for the same reason. The external source jacks accept standard banana plugs with $3 / 4-\mathrm{in}$. spacing.


Figure 1-2, Methods of connection to the measurement terminals.

# Basic Measurement Procedures-Section 2 

2.1 SELECTION OF BRIDGE CIRCUIT ..... $2-1$
2.2 WHEATSTONE BRIDGE MEASUREMENTS ..... $2-2$
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### 2.1 SELECTION OF BRIDGE CIRCUIT

The 1666 Resistance Bridge incorporates 4 bridge circuits to span the range from $1 \mu \Omega$ to $1 \mathrm{p} \mho \mathcal{( 1 T \Omega )}$ with low error introduction. Resistance or conductance measurements can be made with either the Wheatstone or Kelvin bridge circuits. While bath bridges function over the entire range, limits are recommended to preserve accuracy. The bridge type used depends on the approximate value of the unknown. The Wheatstone bridge is not recommended below $100 \Omega$, since lead resistance can cause significant error introduction below this value. The recommended upper limit for Kelvin bridge use is $10 \mathrm{k} \Omega$, since above this value the exposed leads are susceptible to capacitance transients (refer to para 3.3) and leakage resistance to the case, both of which degrade accuracy.

Measurements using the resistance bridges (either Kelvin or Wheatstone) can be made up to $1.1 \mathrm{M} \Omega$. Above this value, conductance must be measured and the reciprocal of the reading made, to obtain the resistance value. Similarly, measurements using either conductance bridge can be made up to 1.1 mho (down to $0.9 \Omega$ ). Above this value, resistance must be measured and conductance obtained from the reciprocal of the reading. Between $0.9 \Omega$ and $1.1 \mathrm{M} \Omega$, either mode and bridge type can be used and readings made directly. Figure 2-1 summarizes the measurement mode and bridge type to be used for a given range i.e., Kelvin-Resistance (KR), Kelvin-Conductance (KG), Wheatstone-Resistance (WR), or Wheatstone-Conductance (WG).


Figure 2-1. Selection of bridge circuit.
1666-2


Figure 2-2. Wheatstone bridge connections,

### 2.2 WHEATSTONE BRIDGE MEASUREMENTS ( $100 \Omega$ - 1.1 M $\Omega$ ).

### 2.2. 1 Connections

Figure 2-2 illustrates the proper methods for connection of the unknown to the bridge for Wheatstone measurements for 2 and 3 -terminals measurements (a), and 2 possible connections when a lead of the unknown must be grounded (b,c).

The unknown resistor is connected between terminals $B$ and $C$ using, either the lead set provided, or other suitable leads, or direct connection to the terminals. Terminals $A$ and D are both guard and should be connected to any guard point required (refer to para 3.6). It is preferable to have the unknown ungrounded and the guard connected to the chassis with the link provided (Figure 2-2a).

The instrument chassis should be well grounded, especially for high-precision comparison measurements or very high resistance measurements, to avoid capacitive pick-up effects. (Refer to para 3.3.).

If one terminal of the unknown must be grounded, connect this terminal to bridge terminal B and either open the link and ground the case (Figure 2-2 b), or connect the link and unground the case (Figure $2-2 \mathrm{c}$ ). The former connection results in lower capacitive transients; however, the latter is less affected by leakage resistance errors.

### 2.2.2 Balance Procedure.

a. Turn the POWER switch to the SOURCE and then the DETECTOR BAT CHECK positions. Observe that the meter reading is in the indicated range in each case (refer to para 1.3 for battery replacement).
b. Turn the POWER switch to INTERNAL BRIDGE SOURCE (refer to para 2.5 for use of external source).
c. Set METER RESPONSE switch to FAST.
d. Set SENSITIVITY control at midrange (dot up).
e. Select for resistance or conductance measurement mode with the R-G switch, located near the RANGE SELECT dial.
f. Set the WHEATSTONE-KELVIN switch to WHEAT. STONE.
g. If the approximate value of the unknown is known, turn the RANGE SELECT dial to the appropriate range. If not, set the ZERO-MEASURE switch to MEASURE, and the 2 left most digits (most significant) of the DECADE to $X X$ and rotate the RANGE SELECT dial (clockwise for the $R$ mode, or counterclockwise for $G$ mode), until the meter deflection reverses.
h. Set the ZERO-MEASURE switch to ZERO and turn the ZERO ADJ control to indicate a reading of 0 on the meter.

1. Set the ZERO-MEASURE switch to MEASURE.
j. Adjust the DECADE digit switches, starting from the left, until the meter indicates 0 . Fast balances can be made by increasing each decade digit until the meter deflection reverses, then "backing off" 1 digit step. If the meter reads to the left of 0 , increase the decade setting; if it is to the right, decrease the setting.

## NOTE

Increase SENSITIVITY, if present setting is inadequate for required accuracy.
k. Set ZERO-MEASURE switch to ZERO and re-adjust ZERO ADJ control, if necessary.
I. If step $k$ resulted in a change, return to the MEASURE position and re-adjust the decade digit switches. If no guard connection is used, the resistance or conductance of the unknown is the decade reading multiplied by the RANGE setting.
m . For guarded connections, depress the BALANCE switch and simultaneously adjust the GUARD OR YOKE control for a zero meter reading. At low decade readings a balance may not be possible, but adjustment as far as possible towards balance should be made.
n. Release the BALANCE switch and re-adjust the decade setting. This setting is the 3 -terminal resistance of the guarded unknown, unless the shunting resistors are below 5 MS (refer to para 4.2.1).

NOTE
For very low readings, greater accuracy is obtained if 1 is added to the least-significant digit, to account for zero resistance of the decade adjustment.

### 2.3 KELVIN BRIDGE MEASUREMENTS $(1 \mu \Omega-10 \mathrm{k} \Omega)$.

### 23.1 Connections.

The unknown is connected to all 4 bridge terminals as illustrated in Figure 2-3a, using either the lead set provided or other suitable leads. The instrument case terminal should be connected to bridge terminal D with the link provided, and these terminals connected to a good ground.

It should be noted that resistance in a 4 -terminal measurement is from the junction of the leads $A$ and $B$ to the junction of leads $C$ and $D$. Terminals $A-D$ are from top-tobottom on the instrument. Thus, if the component has two leads (Figure 2-3b) the resistance measured is that between the two inner connections. If the unknown has two binding posts the resistance of the connections may be avoided by making one connection to a wire through the hole and the other to the top of the binding post. The resistance then measured is that between the two lower (hole) connections (Figure 2-3c). If the unknown has four terminals, one lead should be used to connect to each bridge terminal (Figure 2-3d). In general, connections to $A$ and $B$ are interchangable as are C and D .

### 2.3.2 Balance Procedure.

a. Perform steps a through e of para 2.2.2.
b. Set the WHEATSTONE-KELVIN switch to KELVIN.
c. Perform steps $g$ through 1 of para 2.2.2.
d. If the measurement is on the lowest $R$ or highest $G$ ranges, perform steps $m$ and $n$ of para 2.2.2. (See also para 2.2 .2 note.)

### 2.4 PRECISE COMPARISONS.

Because of the resolution and sensitivity of the 1666, highly accurate comparisons between resistors of nearly equal value can be made. The normal balance procedures are performed but additional care should be exercised, especially with the following:

1. The SENSITIVITY control is set fully cw .
2. The METER RESPONSE is set to NORMAL.
3. Balancing, is performed in both the $Z E R O$ and MEASURE functions.
4. The YOKE balance should be performed on the lowest Kelvin ranges, and the GUARD balance on the highest Wheatstone range, even if a guard is not used. Short heavy-gauge leads should be used for low resistance and conductance measurements. If the unknown has a very high resistance, the leads should be separated to avoid leakage between them.
5. The comparison should be made with as many possible high-order decade digits at the same settings (from left-to-right), since comparison measurements are independent of the accuracy of the unchanged decade digit settings. Therefore, it is good practice to. avoid a setting of $10(X)$ in the high-order digits. If for example, the value of one resistor is $1000.025 \mathrm{k} \Omega$, and the other is $999.975 \mathrm{k} \Omega$, it is preferable to make the first balance at $999 . \times 25$, so that the next balance changes only the last 3 digits. An error of $0.1 \%$ in the fourth highest-order decade results in the same measurement error as a 1 ppm error in the highest-order decade. If this procedure is followed, the error in resistance comparison should be less than $2 \mathrm{ppm}+.001$ (difference in $\mathrm{ppm})$, assuming adequate sensitivity.


Figure 2-3. Kelvin bridge connections.

### 2.5 LIMIT TESTING.

The Type 1666 may be set up to provide a go-no-go indication useful for component testing. The panel meter is used as the indicator. Limit testing with the 1666 is useful when the limits are set no greater than $5 \%$ from the reference used. The setup procedure is as follows:
a. Balance the bridge using a standard resistor decade unit, or a resistor that is preferably within the desired tolerance.
b. Offset the instrument resistance decade by the desired tolerance if the tolerance is symmetrical, or by half the total allowable spread, if unsymmetrical.
c. Adjust the SENSITIVITY control for a 5 -division meter deflection.
d. Set the decade to the center value (the nominal value, if the tolerance is symmetrical).
e. Connect each resistor to be tested to the bridge. If the meter deflection is less than 5 divisions the component is within limits.

When the unknown has a tolerance greater than approximately $\pm 5 \%$, the limits may be in error by more than $1 \%$ if the above method is used. A sure method is to set the decade so that the unknown components at both limits give the same extent of deflection.

## Measurement Considerations - Section 3

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### 3.1 POWER DISSIPATED IN UNKNOWN RESISTOR

The power dissipated in the unknown is a function of the open-circuit source voltage, the resistance of the source, the value of the unknown, the ratio arms, and the bridge circuit used. The actual power at null is determined by a measurement of the voltage across the unknown, or by the following formulas (which indicate the value at null).

Kelvin $R: P_{x}=\frac{\left(E_{b}{ }^{2}\right)\left(R_{x}\right)}{\left[R_{x}+R_{r}+R_{g}+R_{g} R_{r} / 50 \mathrm{k} \Omega\right]^{2}}$
Kelvin $G: P_{x}=\frac{\left(E_{b}{ }^{2}\right)\left(G_{x}\right)}{\left[1+G_{x}\left(R_{r}+R_{g}\right)+R_{g} / 50 \mathrm{k} \Omega\right]^{2}}$
Wheatstone $R: P_{x}=\frac{\left(E_{b}{ }^{2}\right)\left(R_{x}\right)}{\left[R_{x}+R_{r}+R_{g}+R_{g} R_{r} / 8.33 k \Omega\right]^{2}}$
Wheatstone $G: P_{x}=\frac{\left(E_{b}{ }^{2}\right)\left(G_{x}\right)}{\left[1+G_{X}\left(R_{r}+R_{g}\right)+R_{g} / 8.33 \mathrm{k} \Omega\right]^{2}}$
where:
$\mathrm{E}_{b}$ is the open-circuit source voltage (approximately 6 V internal)
$R_{x}$ is the unknown resistor
$R_{r}$ is the ratio-arm resistor
$R_{g}$ is the bridge-source resistor (internal and external)

Table 3-1
RATIO ARM AND BRIDGE-SOURCE RESISTORS FOR EACH RANGE

| Range | Multipliers | $\mathbf{R}_{\mathbf{r}}$ | $\mathbf{R}_{\mathbf{g}}{ }^{*}+$ |
| :---: | :---: | :---: | :---: |
| $100 \mathrm{~m} \Omega$ | 100 mZ | $1 \Omega$ | $62 \Omega$ |
| $1 \Omega$ | 10 mJ | $10 \Omega$ | $182 \Omega$ |
| $10 \Omega$ | 1 mZ | $100 \Omega$ | $492 \Omega$ |
| $100 \Omega$ | $100 \mu \mho$ | $1000 \Omega$ | $882 \Omega$ |
| $1 \mathrm{k} \Omega$ | $10 \mu \mathrm{~J}$ | $10 \mathrm{k} \Omega$ | $1.062 \mathrm{k} \Omega$ |
| $10 \mathrm{k} \Omega$ | $1 \mu \mathrm{~J}$ | $100 \mathrm{k} \Omega$ | $1.062 \mathrm{k} \Omega$ |
| $100 \mathrm{k} \Omega$ | 100 nJ | $1 \mathrm{M} \Omega$ | $1.062 \mathrm{k} \Omega$ |

+This does not include the internal battery resistance, which increases with age.
*When an external source is used, subtract $62 \Omega$ and add any external series resistance.

The values for $R_{r}$ and $R_{g}$ for each range are listed in Table 3-1. Note, when an external source is used, there is no source resistor in the lowest range, and an external resistor should be used to limit the power to 0.5 W (see para 3.1.2).

### 3.1.1 Intemal Source.

The power dissipated by the unknown when the resistance bridges are used is limited by series resistance to less than 0.01 W , to avoid significant errors due to resistor heating. On the $G$ bridge, however, the power dissipated by $R_{x}$ may exceed . 01 W on the lower ranges. The maximum

Table 3-2
POWER DISSIPATED IN UNKNOWN
(For Decade Setting of $1 / 10$ fs On G Bridge With Internal Source)

| Multipliers | $\mathbf{G}_{x}$ | $\mathbf{R}_{x}$ | Power |
| :---: | ---: | :---: | :---: |
| 100 mJ | 100 mJ | $10 \Omega$ | .068 W |
| 10 mJ | 10 mJ | $100 \Omega$ | .042 W |
| 1 mJ | 1 mz | $1 \mathrm{k} \Omega$ | .014 W |
| $100 \mu \mho$ | $100 \mu \mho$ | $10 \mathrm{k} \Omega$ | .0025 W |
| $10 \mu \mho$ | $10 \mu \mho$ | $100 \mathrm{k} \Omega$ | $290 \mu \mathrm{~W}$ |
| $1 \mu \mho$ | $1 \mu \mho$ | $1 \mathrm{M} \Omega$ | $30 \mu \mathrm{~W}$ |
| 100 nv | 100 nv | $10 \mathrm{M} \Omega$ | $3 \mu \mathrm{~W}$ |
|  |  |  |  |

power is about 0.15 W when $.016 \mho(63$ ohms) is measured on the lowest range. However, this is a setting below $1 / 10$ th full scale on the lowest decade readout and would normally not be used. The power for measurements on the G bridge at $1 / 10$ th full scale is given in Table 3-2. Power dissipation decreases at higher settings.

### 3.1.2 External Source.

The power dissipated in the unknown when an external source is used, can be calculated from the formulas given and from the constants in Table 3-1. Note that when an external source is used the $R_{g}$ value of Table 3-1 should be reduced by $62 \Omega$ and increased by any external resistance used. It is desirable to limit the applied power to 0.5 W or less.

### 3.2 THERMAL VOLTAGES.

A closed circuit consisting of conductors of different materials will have a current flowing in it, if the junctions of these conductors are not at the same temperature. This is the Siebeck effect. The voltage produced is approximately proportional to the temperature difference between the junctions and a constant that depends on the materials used. This constant is only a few $\mu \mathrm{V} /{ }^{\circ} \mathrm{C}$ for combinations of various copper alloys, but can be much higher for some combinations of metals.

Because most resistors have leads of some alloy of copper, the 1666 Bridge has solid copper alloy binding posts and the lead set has copper alligator clips and banana pins to keep thermal voltages low. Gold plating is used to avoid corrosion; there is negligible thermal voltage, since the plating is so thin it can have only a small thermal gradient. The main cause for a thermal gradient is a result of body heat when the binding posts are tightened or the alligator clips handled. Also friction in plugging-in the banana pins of the lead set can result in a slight temperature change.

The effect of thermal voltages is greatest when the bridge voltage is low as it is when measuring low-valued resistors on the lowest range. A resistance change of $1 \mu \Omega$ produces a $0.1 \mu \mathrm{~V}$ change in output voltage. Thus, small thermal
voltages can cause appreciable errors. However, in most cases the temperature difference decreases quickly.

Thermal voltages affect the detector indication equally for both the MEASURE and ZERO functions. Therefore, a constant thermal voltage will give no error if the detector is zeroed. However, because the thermal voltage will usually disappear quickly, it is preferable to wait until the detector indicator is steady before making low ohm measurements.

### 3.3 CAPACITANCE TRANSIENTS.

When measuring high resistance values, currents due to capacitive coupling to exposed leads can cause large meter transients. These currents may be due to a fixed capacitance to a changing voltage or a changing capacitance to a fixed voltage.

These transients can usually be removed by: 1. grounding the bridge case with terminal $D$ connected to the case terminal, 2. grounding the operator, 3. grounding nearby equipment, 4. using short leads.

In extreme cases such as when long leads must be used to connect high impedance resistors, the leads should be shielded, with the shields connected to the grounded bridge case.

### 3.4 A-C PICKUP.

The detector input filter and the selective a-c amplifier stages greatly attenuate the effect of power-line hum and other a-c pickup. The main effect of pickup is loss of sensitivity. The detector is relatively sensitive to pickup near the modulator frequency, about 750 Hz , or its harmonics. Pickup synchronous to the frequency will cause a constant bridge deflection.

The worst condition is pickup to the " $C$ " terminal when high resistances are measured on the Wheatstone bridge. $A$ $60-\mathrm{Hz}$ voltage of 3 V , coupled to the " C " terminal by 100 pF , will make a $.01 \%$ measurement of a $1 \mathrm{M} \Omega$ resistor difficult, because of the resulting loss of detector sensitivity.

Pickup can be removed by the same precautions used to avoid capacitance transients (See para 3.3.).

### 3.5 MEASUREMENTS UNDER HUMID CONDITIONS.

High humidity results in an increase in the conductivity of insulators and can even produce electrolytic voltages, if contaminating substances are present. The combination of humidity and high temperatures is particularly troublesome. While the 1666 bridge uses materials and components chosen for good operation at high humidity and uses internal guarding to reduce the effects of leakage currents, at very high humidity and high temperatures, the accuracy of the bridge is reduced, particularly when making high resistance measurements. While the best accuracy can be expected only over a low temperature range at low humidity, a $0.1 \%$ accuracy should be possible over the


Figure 3-1. Bridge accuracy versus temperature/humidity.
ranges of humidity and temperature illustrated in Figure 3-1, as long as the bridge has not been exposed to excessive contamination.

Excessive humidity-temperature effects usually first appear as an inability to zero the instrument when the detector offset range, approximately $\pm 100 \mu \mathrm{~V}$, is not adequate to counteract the effects of leakage currents. Generally, if the bridge can be zeroed, the humidity effects are not serious.

Several precautions aid bridge operation at high humidity:

1. Connect the " $D$ " terminal to the case terminal with the link provided.
2. Make sure external leads are not touching each other.
3. Set the first digit of the readout to zero.
4. Heat the bridge case slightly with a radiant lamp, or a contacting source of heat, to quickly reduce the internal relative humidity.

It is important to remember that the unknown resistance itself is subject to humidity effects, and leakage across it can produce what can appear as a bridge error.

### 3.6 RESISTIVITY MEASUREMENTS.

Resistivity measurements generally fall into 2 categories: measurements on low-resistance conductors and measurements on high-resistance insulators; however, there is also an in-between range of medium-resistivity materials, particularly semi-conductors. These categories require 3 - or 4 terminal measurements, while in the middle range simple 2-terminal measurements are adequate. The 1666 bridge performs all of these types of measurements within its ranges and specifications.

The main difficulty with resistivity measurements is the conversion of resistance measurements into resistivity, since the conversion depends on the geometry of the specimen. This procedure has been standardized by the American Society For Testing Materials and a partial list of resistivity standards is given in Table 3-3. Refer to the ASTM Standards Index for a complete list of test configurations and dimensions.

Table $3-3$

| SOME ASTM RESISTIVI Title | STANDA Std | V Vol. |
| :---: | :---: | :---: |
| DC Resistance or Conductance of Insulating Materials | D257 | 27.28 or 29 |
| Resistivity of Electric Conductor Materials | B193 | 5,6 or 8 |
| Resistivity of Semi-conductor Materials | F43 | 8 |
| Resistivity of Siticon Slices with a |  |  |
| Collinear Four Probe Array | F84 | 8 |
| Specific Resistance of Electric |  |  |
| Insulating Liquids | 01169 | 29 |

### 3.7 USE OF AN EXTERNAL SUPPLY.

An external supply is useful for increasing sensitivity when high resistance measurements are made. Sensitivity is proportional to the voltage supplied. A maximum of 100 V may be applied to the 1666 , but it is desirable to limit the voltage to 30 V for safety reasons; 30 V is adequate for nearly all measurements. The input power should be limited to 0.5 W . Power limiting is most easily accomplished by insertion of a $0.5-\mathrm{W}$ resistor of $0.5 \mathrm{E}^{2} \Omega$ in series with the external source. Note that, with more power applied to the bridge, resistance changes increase because of power dissipation. For high resistance measurements (greater than $100 \mathrm{k} \Omega$ ) the main bridge standard ( $50 \mathrm{k} \Omega$ ) receives the most power. This is a maximum of 0.2 W , if 100 V is applied, and could result in a 20 ppm change in its value.

Sensitivity can be increased for low resistance measurements by using an external power supply that will provide more current capability than the 0.1 A furnished by the internal supply. The power applied should be limited to 0.5 W .

NOTE
Increased power to the bridge can result in error because of the power dissipation in the ratio arm. A dissipation of 0.5 W could cause an error of $.04 \%$.

## Theory-Section 4

4.1 GENERAL ..... 4-1
4.2 WHEATSTONE BRIDGE ..... $4-1$
4.3 KELVIN BRIDGE ..... 4.3
4.4 BRIDGE SOURCE ..... 4.4
4.5 BRIDGE DETECTOR . ..... 4.4

### 4.1 GENERAL.

The 1666 contains 4 bridge circuits: a guarded Wheatstone resistance bridge, a guarded Wheatstone conductance bridge, a Kelvin (Thomson) resistance bridge, and a Kelvin (Thomson) conductance bridge. The main circuit elements are common to all bridges, but are switched to the various configurations by the WHEATSTONE-KELVIN and R-G switches. The various bridge circuits are all variations of the Wheatstone Bridge.

### 4.2 WHEATSTONE BRIDGE.

Figure 4-1 illustrates the basic Wheatstone bridge. For this circuit the output voltage is zero or "at null" if $R_{X} / R_{A}=$ $R_{r} / R_{B}$ or $R_{X}=R_{A}\left(R_{r} / R_{B}\right)$. In the resistance bridges, $R_{A}$ is a decade resistor ( $55.5555 \mathrm{k} \Omega$ max.). The resistance of the decade is directly proportional to the value $R_{x}$. The value or $R_{r}$ (the ratio arm) is changed with the range switch in in decade values from $1 \Omega$ to $1 \mathrm{M} \Omega$.

The above equation may also be expressed as $1 / R_{X}=$ $G_{X}=R_{B} / R_{A} R_{r}$ for conductance measurements. In the conductance bridges, $R_{B}$ becomes the adjustable decade and the conductance of the unknown is then proportional to $R_{B}$ and its digital readout.

The Wheatstone bridge has inherent advantages over other resistance measurement methods that have made it useful for over 100 years. Some advantages are:

1. The unknown is compared against precision resistors, which can be accurate and stable.
2. The null condition is independent of applied voltage and the source or detector impedance; these affect only the sensitivity.
3. The null point is independent of detector linearity.

### 4.2.1 Three-Terminal or "Guarded" Resistors.

When high-valued resistors are measured, the unknown is often part of a 3-terminal network, as shown in Figure 4-2. Here $R_{X}$ is the value to be measured and $R_{p}$ and $R_{Q}$ are
shunting resistances to a third (or guard) terminal G. While $R_{P}$ and $R_{Q}$ may be actual resistors, more often they are leakage resistances to shields or other circuit points. When measuring the resistivity of high-resistance material, they represent the resistance to the guard electrode (refer to paragraph 3.6.) They can also represent leakage resistance between the unknown terminals of the bridge and the bridge case.


Figure 4-1. Basic Wheatstone bridge.


Figure 4-2. Three-terminal network.

(a)


1666-8

Most high-resistance standards are shielded, with the case as the guard terminal. Some very high-valued standards are actually $T$-networks, which can be converted to a $\Delta$ configuration by the $Y-\Delta$ transformation. This is illustrated in Figure 4-3. If $R_{2}$ is low, the effective value of $R_{X}$ can be extremely large.

While the value of $R_{X}$ in Figure 4.2 may be determined by 3 separate 2 -terminal measurements, it is more easily measured if the bridge circuit used is immune from the shunt loading of $R_{p}$ and $R_{Q}$. There may be negligible errors with the conventional Wheatstone bridge if the guard point (Figure 4-2, point $G$ ) is tied to the bridge point $C$ or $D$ in Figure 4-1. If this is done, either $R_{p}$ or $R_{Q}$ shunts either $R_{r}$ or $R_{A}$. The error is negligible only if the shunted resistor is sufficiently low.

### 4.2.2 The Guard Circuit.

If 2 bridge arms are added to the basic Wheatstone bridge, and their junction is used as a guard point, 3 -terminal measurements can be made with 3 -terminal resistors with no error. This addition to the bridge circuit is called a Wagner guard circuit. The bridge configuration is shown in Figure 4-4. The condition for zero error occurs if $R_{X} / R_{r}=$ $R_{A} / R_{B}=R_{a}^{\prime} / R_{b}$ where $R_{a}^{\prime}$ is the parallel combination of $R_{a}$ and $R_{p}$. The amount of error depends on $R_{Q}$ and the difference between $R_{A} / R_{B}$ and $R_{a}^{\prime} / R_{b}$, i.e.

$$
\% \text { error } \simeq \frac{R_{r}}{R_{Q}}\left(\frac{R_{A}}{R_{B}}-\frac{R_{a}^{\prime}}{R_{b}}\right) \frac{R_{b}}{R_{a}^{\prime}+R_{b}}
$$

In the $1666, R_{a}$ and $R_{A}$ are ganged decades, so that $R_{A} / R_{B} \simeq R_{a}^{\prime} / R_{b}$. This balance is not perfect because of bridge resistor tolerances and because of the effect of $R_{p}$ on $R_{a}^{\prime}$. Therefore, the GUARD-YOKE balance provides a slight adjustment on the $R_{a}^{\prime} / R_{b}$ ratio to make the error negligible.

### 4.2.3 Four-Terminal Measurements.

When a low-valued resistor is measured on a 2 -terminal bridge, the resistance of the bridge terminals, the leads, and the contacts are added to the unknown measured. It is possible to correct for some of the resulting error by making a second measurement with the leads shorted and subtracting their measured resistance from the measurement of the unknown. However, this technique does not
remove errors from differences in contact resistance for the 2 measurements.

In 4-terminal measurements, 4 leads are used to avoid the effect of lead and contact resistance. In the ideal case, this is a transfer-impedance measurement, where current is applied to 2 terminals of a network (Figure 4-5), and the resulting voltage is measured at two other terminals. The transfer of 4 -terminal resistance is defined as the open-circuit output voltage divided by the input current. This ratio is $R_{X}$ in Figure $4-5$ and is completely independent of the resistance in each lead. This is the basis for 4 -terminal bridges, although they are not quite as simple as this ideal case. In such bridges, the lead impedances have some effect that should be considered, even though it may be small.

$1666 \cdot 9$

Figure 44. Wheatstone bridge with Wagner guard circuit.
$*$


1666-10
Figure 4-5. Hlustration of Impedance Transfer.

Low-valued resistance standards generally have 4 terminals, so that any good 4 -terminal measurement will give the same value independent of (external or internal) lead or contact resistance.

### 4.3 THE KELVIN BRIDGE

A 4-terminal resistor can be measured on a Wheatstone bridge having 4 terminals, as shown in Figure 4-6. Resistances $r_{1}$ and $r_{4}$ introduce no error (but may reduce sensitivity); however, $r_{2}$ and $r_{3}$ introduce errors of $-r_{2} / R_{A} X$ $100 \%$ and $-r_{3} / R_{r} \times 100 \%$. This is preferable to the 2-terminal error $\left(r_{2}+r_{3}\right) R_{X} \times 100 \%$ only if $R_{r}$ and $R_{A}$ are $\gg R_{X}$. Generally, the adjustable arm, $R_{A}$, is relatively large, so that $r_{2}$ causes only a small error. However, the ratio arm, $R_{r}$, is generally low-valued, when low resistances are measured, so that the error $r_{3} / R_{r} \times 100 \%$ may be large.

The $\mathbf{r}_{3}$ error is caused by having the voltage drop across $r_{3}, e_{g}$, added to the voltage across $R_{r}$. If the detector were connected to point $B$, this drop would add to that of the $R_{X}$ arm, giving an error of $+r_{3} / R_{X} \times 100 \%$. If voltage drop $e_{g}$ could be divided between the $R_{X}$ and $R_{r}$ arms in the correct proportion, the error could be zero.

The Kelvin bridge, Figure 4-7, contains an additional pair of bridge arms, $R_{a}$ and $R_{b}$, to divide $e_{g}$ between the two sides of the bridge. The measured value of resistance is now (approx)

Rmeas $=R_{X}-\frac{r_{2} R_{r}}{R_{B}}+\frac{r_{4} R_{b}}{R_{a}+R_{b}}\left(\frac{R_{a}+r_{3}}{R_{b}}-\frac{R_{A}}{R_{B}}\right)$
The percent of error due to first term is $r_{2} R_{r} / R_{X} R_{B} X$ $100 \%=r_{2} / R_{A} \times 100 \%$ and is the same for all ranges.

Figure 4-6. Basic Kelvin Bridge.


In the $1666, R_{A}$ is $50 \mathrm{k} \Omega$ at full setting, with least significant digits of . $05 \Omega$ per step. Thus, a lead resistance of, say, $0.1 \Omega$ results in a measurement that is low by 2 steps of the last decade. Such an error is generally negligible, except for comparison measurements, in which case it can be made negligible if $r_{2}$ has approximately the same value for both measurements.

The last error term is made small by ganging $R_{a}$ with $R_{A}$ so that the ratio $R_{a} / R_{b} \simeq R_{A} / R_{B}$ and made even smaller by the YOKE adjustment that trims the $R_{a} / R_{b}$ ratio. With no yoke adjustment, the value of $R_{b} /\left(R_{a}+R_{b}\right)\left[R_{a}+\left(r_{3} / R_{B}\right)\right.$ - $\left.R_{A} / R_{B}\right]$ for the 1666 is typically less than .002 at full scale, and much less at lower settings, even if $r_{3}=1 \Omega$. Therefore, a lead resistance, $\mathrm{r}_{4}$, of $0.1 \Omega$ generates an error of only $200 \mu \Omega$ or $.02 \%$ at full scale on the lowest $R$ range. At low settings, it would be less than $10 \mu \Omega$. On higher bridge ranges this error is negligible except for precise comparisons.

The yoke adjustment is made by increasing $r_{4}$ by $10 \Omega$ and adjusting the yoke control to balance. This reduces the error by a factor of 100 , making it negligible.

The yoke adjustment can modify $R_{b}$ by about $\pm 0.3 \%$ and add $0.2 \Omega$ to $R_{2}$. Resistance cannot be subtracted from $R_{a}$ to compensate for lead resistance (but the last decade of the adjustment does not add resistance in $R_{a}$, so it can be $0.1 \Omega$ low). Therefore, yoke balances cannot be made at many low decade settings. However, if the lead resistances ( $r_{3}$ and $r_{4}$ ) are $0.1 \Omega$ or less, the resulting measurement error is $1 \mu \Omega$ or less, if the best possible yoke balance is made (full cW ).


Figure 4-7. Kefivin bridge with yoke adjustment.

### 4.4 THE BRIDGE SOURCE.

The bridge source consists of four 1.5 V-D cells in series with a series resistance of $62 \Omega$ to limit the current to less than 0.1 A . On all but the lowest resistance range, additional resistance is placed in series with the source to limit the applied power. The source voltage is checked with a load of $200 \Omega$ when the POWER switch is set to the SOURCE BAT CHECK position.

The polarity of the source is switched as the various bridges are selected to insure a right-hand meter deflection when the decade setting is too high.

### 4.5 THE BRIDGE DETECTOR CIRCUITS. (Figure 6-4)

### 4.5.1 General.

Figure $4-8$ is a block diagram of the detector circuits. Section 6 contains the schematic diagrams of these circuits. The detector consists of an input low-pass filter, an FET modulator-chopper, a tuned 3-stage ac amplifier, FET demodulator, an RC Wien-Bridge oscillator, and a meter response filter network. Refer to the detector specifications at the front of the manual. The following paragraphs describe the detector circuits.

### 4.5.2 Detector Power.

The detector is powered by four $1.5-\mathrm{V}$ D cells in series. The voltage is checked when the POWER switch is set to the DETECTOR BAT CHECK position.

### 4.5.3 Input Filter.

The input filter is low-pass and consists of C1 and the bridge output resistance network. It reduces the effect of hum and other ac pick-up. Parallel-opposed diodes CR1 and CR2 limit the input voltage to prevent large dc overloads.

### 4.5.4 Modulator.

The FET modulator (Q1) modulates the filtered dc input signal from the bridge with a low-level sinewave (approx. 750 Hz ) from the oscillator. The modulator out put is fed to the amplifiers (para 4.5.6). There are internal
adjustments (R40 and R51) for FET bias adjustment and to cancel the effect of capacitance across the FET. The modulator signal changes the resistance of the FET (Q1), so that the output of the divider formed by R1 and Q1 has an ac component when dc is applied to the input.

### 4.5.5 Zero Set.

The zero-adjust control (A.R2) inserts an in phase a-c signal of either polarity to compensate for input voltage offset of $\pm 100 \mu \mathrm{~V}$. A-R2 has an 11-turn range to provide high resolution.

### 4.5.6 Amplifier.

The Modulator output is applied to 3 cascaded high-gain, high-Q selective amplifiers ( $\mathrm{Q} 2-\mathrm{Q} 9$ ), which reject noise that could otherwise overload the detector.

### 4.5.7 Sensitivity Control.

The sensitivity control (A-R3) is a logarithmic potentiometer that is located between amplifier stages 1 and 2. The control allows adjustment of the gain depending on the input conditions.

### 4.5.8 Demodulator.

The FET demodulator (Q14) converts the a-c amplifier output to dc. The oscillator output through Q13 and turns FET Q14 on and off to give synchronous detection.

### 4.5.9 Meter-Response Filters.

The meter-response filters consist of 2 cascaded R-C ladder sections, in which capacitor values are switched in for the different response-speed positions. Pairs of parallelopposed diodes bridge (CR12, CR13) and load (CR2, CR6) this network. The loading pair of diodes prevents excessive output capacitor voltage and meter current. The bridging pair of diodes conducts only at high output levels and effectively by-passes the filter for large unbalances. Thus, the meter response is relatively slow for small unbalances, as it is desirable to reduce noise, but for large unbalances it provides for fast coarse adjustment.


Figure 4-8. Block diagram of the 1666 detector circuits.

## Service and Maintenance-Section 5

5.1 GR FIELD SERVICE ..... 5.1
5.2 MINIMUM PERFORMANCE STANDARDS ..... 5-1
5.3 CALIBRATION CHECKS ..... $5-2$
5.4 TROUBLE ANALYSIS ..... 5.3
5.5 CALIBRATION PROCEDURE ..... $5 \cdot 3$
5.6 KNOB REMOVAL/INSTALLATION ..... $5 \cdot 6$

### 5.1 GR FIELD SERVICE.

The warranty attests the quality of materials and workmanship in our products. When difficulties do occur, our service engineers. will assist in any way possible. If the difficulty cannot be eliminated by use of the following service instructions, please write or phone our Service Department (see last page of manual), giving full information of the trouble and of steps taken to remedy it. Be sure to mention the serial, ID, and type numbers of the instrument.

### 5.2 MINIMUM PERFORMANCE STANDARDS.

The 9 checks listed in Table 5-1 are given so that it can be determined that the instrument is in proper working condition (1) on receipt of a new bridge, (2) after a period of non-use, or (3) after repairs have been made to the bridge. If any specifications (Read Column) are not met, it is likely that 1 or more potentiometers are misadjusted. The calibration procedure should be performed until the misadjustment is corrected. Measurements should be made

Table 5-1
ACCURACY AND OPERATIONAL CHECKS

| GR Cat. No. | Value | Parameter Switches* | Multiplier | Read** |
| :---: | :---: | :---: | :---: | :---: |
| 1440-9601 | $1 \Omega$ | KR | 100 ms | $R=X .00000 \pm 0.011 \%$ |
| 1440-9611 | $10 \Omega$ | KR | $1 \Omega$ | $R=X .00000 \pm 0.011 \%$ |
| 1440-9621 | $100 \Omega$ | KR | $10 \Omega$ | $R=\times .00000 \pm 0.011 \%$ |
| 1440-9631 | $1 \mathrm{k} \Omega$ | KR | $100 \Omega$ | $R=\times .00000 \pm 0.011 \%$ |
| 1440.9631 | $1 \mathrm{k} \Omega$ | WR | $100 \Omega$ | $R=X .00000 \pm 0.011 \%$ |
| 1440-9631 | $1 \mathrm{k} \Omega$ | WG | $100 \mu \Omega$ | $\mathrm{G}=\mathrm{X} .00000 \pm 0.011 \%$ |
| 1440-9641 | $10 \mathrm{k} \Omega$ | WR | $1 \mathrm{k} \Omega$ | $R=X .00000 \pm 0.011 \%$ |
| 1440-9651 | $100 \mathrm{k} \Omega$ | WR | $10 \mathrm{k} \Omega$ | $R=X .00000 \pm 0.011 \%$ |
| 1440-9661 | $1 \mathrm{M} \Omega$ | WR | $100 \mathrm{k} \Omega$ | $R=\times .00000 \pm 0.011 \$$ |

[^0]Table 5-2
RECOMMENDED TEST EQUIPMENT*

| Name | Minimum Use Specification | Recommended Equipment** |
| :---: | :---: | :---: |
| Oscilloscope | Vertical Sensitivity: $=50 \mathrm{mV} / \mathrm{cm}$ <br> Range: 100 kHz | Tektronix Model 503 |
| Decade Resistor | 6 decade, $1 \Omega$ per step | GR Type 1433-F, G, or $X$ |
| Frequency |  |  |
| Meter Range: | Range: $735 \mathrm{~Hz}-765 \mathrm{~Hz}$ | GR Type^1192 Counter |
| Standard | Value Accuracy*** |  |
| Resistors | $1 \Omega \quad \pm 0.02 \%$ | GR Type 1440-9601 |
|  | $10 \Omega \pm 0.01 \%$ | GR Type 1440-9611 |
|  | $100 \Omega \quad \pm 0.01 \%$ | GR Type 1440-9621 |
|  | $1 \mathrm{k} \Omega \quad \pm 0.01 \%$ | GR Type 1440-9631 |
|  | $10 \mathrm{k} \Omega$ : $\pm 0.01 \%$ | GR Type 1440-9641 |
|  | $100 \mathrm{k} \Omega \quad \pm 0.01 \%$ | GR Type 1440-9651 |
|  | $1 \mathrm{M} \Omega \quad \pm 0.01 \%$ | GR Type 1440-9661 |
| Microvolt Source | $\pm 2 \mu \mathrm{Vdc}$ | GR Type 1346 Audio-Frequency Microvoltmeter ${ }^{\text {B }}$ |
| Cable Assy. | 1666-4 Lead, Molded | GR Type 1666-2200 (suppiled) |

* Instruments recommended for minimum-performance standards and trouble analysis.
* Or equivalent.
**Calibrated to 30 ppr .
between $20^{\circ} \mathrm{C}$ and $25^{\circ} \mathrm{C}$ at less than $75 \% \mathrm{RH}$ (see specifications). Table 5-2 lists the recommended test equipment for these checks plus the equipment needed for the calibration procedures given later. Refer to para 2.2 and 2.3 for the balancing procedures.

The proper method for connecting the bridge to a 1440 for a Kelvin connection using the lead set provided, is to connect lead A to one post of the 1440; lead B to the banana plug (bottom of the same post); and connect leads $C$ and $D$ in a similar manner with the other 1440 post and plug.

### 5.3 CALBBRATION CHECKS

### 5.3.1 Resistance Decade (RA).

a. Connect a GR Type 1433-F, -G, or -X Decade Resistor to the 1666 Bridge as shown in Figure 5-1. Remove the 1433 ground link.
b. Perform Source and Detector BAT CHECKS.


Figure 5-1. 1666/1433 Connections using Cable Assembly 1666-2200.
c. Set the 1666 controls as follows:

POWER - INTERNAL BRIDGE SOURCE
METER RESPONSE - FAST
SENSITIVITY - MIDRANGE
R/G-R
RANGE SELECT - $\times 10 \mathrm{~K}$
WHEATSTONE/KELVIN - WHEATSTONE ZERO/MEASURE - ZERO
(adjust ZERO ADJ control to indicate reading of 0 on meter)
ZERO/MEASURE - MEASURE
1666 DECADE RESISTOR $-\times .00000$
d. Adjust the 1433 for a balance as indicated by the 1666 front-panel meter. Record the 1433 reading.
e. Lower the most significant 1666 decade lever switch from $X$ to 9 and raise the value of the 2 nd most significant decade switch to $X$.
f. Rebalance the 1433. The difference should not be more than $2 \Omega$ ( 20 ppm ).
g. Repeat the procedure (lower the $X$ to 9 , raise the next most significant decade lever to $X$ ) for each decade lever switch. Each 1433 reading should not vary more than $2 \Omega$ for the value initially recorded.

### 5.3.2 Sensitivity.

With a balance achieved during the decade-resistor calibration check in the previous paragraph, and no other changes made to the test setup, vary the right-hand-most 1666 decade lever-switch (LSD) 2 digits from its present setting. The meter should deflect approximately 1 division.

### 5.3.3 Meter Response.

Using the test setup described in para. 5.3.1, change the decade step that produces approximately a +5 to -5 meter movement at FAST (usually the 3rd decade from the right). Repeat the procedure with the METER RESPONSE set to NORMAL, and then SLOW. The NORMAL response should be observed as approximately half the speed of FAST, and SLOW should be approximately half the speed of NORMAL.

### 5.4 TROUBLE ANALYSIS.

Figures 6-2,6-3

### 5.4.1 Preliminary Checks.

If satisfactory measurements are difficult or impossible to obtain, make the following external checks first:

1. Is the unknown connected correctly?
2. Are the panel controls set correctly?
3. Are the batteries correctly in place and are both battery checks successful?

### 5.4.2 Measurement Errors.

Measurement errors due to faulty or misadjusted bridgecircuit components can be located by means of the resistance checks of Table 5-1.

## Symptoms and Probable Remedies

1. When measurements on a particular range are in error, the ratio arm $R_{R}$, which consists of a potentiometer and fixed resistors in a series-parallel combination (refer to switch assembly AS4-AS8 schematic diagram Figure 6-5) is in error. Perform the calibration procedures for these adjustments (R101, R103, R105, R107, R109, R110 and R111) before replacing any ratio-arm resistors.
2. When all measurements are in error, adjustment of R112 (the standard arm adjustable resistor) should be performed. See para 5.5.7. Verify the value of R161 if the adjustment of R112 is out of range.
3. If the measurements are still in error perform the decade check (para 5.3.1).

### 5.4.3 No Meter Indication.

No meter indication, or a low meter indication may be due to weak or dead batteries, no oscillator output, poor detector sensitivity, faulty meter-response filter, or a faulty meter. If the batteries are not at fault, perform the detector calibration procedures to aid in fault isolation, and the procedure in para 5.4.4.

### 5.4.4 No Meter Deflection With A Large D-c Input (1 V) Applied.

a. Observe: TP 2 should be approximately 3.0 V ac pk-to-pk. If it is not, check Q15 and Q16.
b. Observe: TP 3 should be a squarewave with a lower value of approximately -0.5 V and an upper value of approximately +1.5 V .

1. If the signal drops to a relatively large negative voltage check the gate drive of 014 (approx. $\pm 3 \mathrm{~V}$ squarewave), and Q13 input ( $\pm 0.4 \mathrm{~V}$ squarewave).
2. If the signal is correct, check the meter leads, CR6 and CR7, and the meter (BRIDGE SOURCE) switch. Continue if no signal.
c. No signal at TP 3.
3. Check AT4 for a $3 V$ ac $p k-p k$ (approx.) signal.
4. If the signal is present at AT4 but not at AT6, insure that the SENSITIVITY control is set fully cW .
5. Check amplifier stages Q 5 to Q 12 (see para 5.5.6).

Continue if there is no signal.
d. No Signal At AT4 (TP1).

1. Check that the signal at AT2 is approximately 200 mV ac pk-pk.
2. If the signal is present, then check $\mathrm{Q} 2, \mathrm{Q} 3$, and Q 4 . Continue if there is no signal.
e. No Signal At AT2
3. Check the gate drive of 01 for a 1.0 V ac $p k$-pk (approx.) signal.
4. If the signal is absent check Q1.
5. Continue if there is still no signal.
f. No Signal At Q1 Gate.
6. Observe whether there is a signal at AT12 of 1.0 V ac pk-pk (approx.)
7. If the signal is not present, check the circuit between the oscillator and AT12.

### 5.5 CALIBRATION PROCEDURES.

### 5.5.1 General.

The few internal adjustments are factory set and normally do not require readjustment. Procedures for readjustment are included here but should be used only when the operator is reasonably certain that readjustment is necessary.

### 5.5.2 Equipment Required.

The equipment necessary to perform the following calibration procedures is listed in Table 5-2.

### 5.5.3 Etched-Board Access.

For access to the bridge board shown in Figure 5-2 perform the following:
a. Open the instrument cabinet until the instrument and cover are at $90^{\circ}$.
b. Remove the 2 cabinet screws located on the rear of the cabinet.
c. Lift the instrument from the cabinet and place it on either side.
d. Locate and remove the two interior and board retaining screws, as shown in Figure 5-2 on each side.
e. The back plate holding the battery tubes is now free to swing up and away from the interior. Both sides of the


Figure 5-2. 1666 Bridge board adjustments (1666-4700)


Figure 5-3. 1666 rear interior.
bridge board are now accessible for the calibration procedures.

### 5.5.4 Oscillator, Sensitivity Controls - Initial Setup.(Fig. 5-2)

a. Set the POWER control to INTERNAL BRIDGE SOURCE. Turn the ZERO ADJ control $51 / 2$ turns to center (11-turn potentiometer): SENSITIVITY control to MID-RANGE. Additionally set the 1666 to WHEATSTONE, R, MULT BY $100 \mathrm{~K} \Omega$, FAST.
b. Prepare the oscilloscope as follows:

Sensitivity - $50 \mathrm{mV} / \mathrm{cm}$
Input-AC
Sweep Time/cm - 0.2 ms
Trigger - External
c. Ground the oscilloscope to the bridge board; place the sync on TP2 and connect the signal probe to C13 (high).
d. Prepare the AF Microvolter as follows:

OUTPUT - OFF
METER FULL SCALE - $\pm 10 \mathrm{Vdc}$. FULL SCALE OUTPUT VOLTAGE - $10 \mu \mathrm{~V}$
e. Connect the Microvolter OUTPUT (high) to the bridge board anchor terminal AT1; the low output connects to AT3. Use a short shielded lead.
f. Adjust the counter for frequency measurement of approximately 750 Hz , and adjust the gate time to 1 . Connect the counter input to TP2 and the shield strap to bridge board ground.

### 5.5.5 Oscillator Adjustment.

(Figure 5-2)
a. Turn the bridge-board BIAS R47 for maximum sinewave amplitude.
b. Adjust the bridge front-panel SENSITIVITY control (approx. center of potentiometer) to produce an undistorted sinewave.
c. Adjust bridge-board FREQ R40 for a maximum amplitude on the oscilloscope. (This tunes the oscillator to the 3-stage amplifier.)
d. Observe with the frequency meter that the frequency is between 735 Hz and 765 Hz . If the frequency is outside of this range, one or more of the 3 amplifier stages is out of tolerance (refer to para 5.5.6).
e. Adjust C-BAL R51 for minimum amplitude on the oscilloscope.

### 5.5.6 Sensitivity Adjustment.

(Figures 5-2, 6-2)
a. Connect oscilloscope probe to TP3.
b. Increase 1666 front-panel SENSITIVITY control to maximum cW . The waveform described in para 5.5 .5 is obtained by the proper adjustment of the C-BAL control and the front-panel ZERO ADJ control. The ZERO ADJ control should be approximately $\pm 1$ turn from the center position.
c. Rotate the ZERO ADJ control and the front-panel meter must move in the same direction (to the right for a ow turn).
d. Turn on the Microvolter and adjust the level for $1 \mu \mathrm{~V}$ output.
e. Move the METER FULL SCALE control +10 Vdc to -10 Vdc . The front-panel meter should move approximately 4 divisions ( $\pm 2$ divisions).
f. If the sensitivity is not enough, readjust BIAS R47 in small increments, rezero, and check with the Microvolter. If the sensitivity is still low readjust the C-BAL, R51, in the same manner.
g. Check the gain of the ampifier (TP2) for a 3 V ac pk-pk (approx.) signal when there is a 0.1 Vdc input. If $T P 2$ is low, check the first stage ( 01 in particular). If it is satisfactory continue with step $h$.
h. Check at the junction of C 7 and C 8 for a 4 V ac $\mathrm{pk}-\mathrm{pk}$ (approx.) signal, when there is 1 mV dc input. If the junction voltage is low check the second stage. If it is satisfactory continue with step $i$.
i. Check the junction voltage at C 13 and R29 for a 4 V ac pk-pk (approx.) signal when there is a $10 \mu \mathrm{~V}$ dc input. If the junction voltage is low check the third stage.
j. Readjust R47 (step f) as necessary.

### 5.5.7 Bridge Adjustments.

## Standard Arm

a. Set the 1666 front-panel controls as follows:

BRIDGE SOURCE - Internal
WHEATSTONE/KELVIN - WHEATSTONE
R/G-R
SENSITIVITY $-1 / 3 \mathrm{cw}$
ZERO/MEASURE - MEASURE (perform zero adjust first.)
MULTBY - $100 \Omega$
METER RESPONSE - FAST
DECADE - $999 \times 00$
b. Set R107 and R112 to mid-range.
c. Plug into the 1666 unknown terminals a GR Type $14401-k \Omega$ resistor, and adjust $R 107$ for null on the meter.
d. Set the R/G Switch to G and adjust R112 for a null on the meter. The two operations interact with each other.
e. Repeat the R107 and R112 adjustment procedures in sequence with increasing sensitivity, until the SENSITIVITY control is at maximum cW , and the change in R107 and R112 is reduced to 1 division. Convergence is simplified if underadjustment is performed i.e., adjustment less than required for null.
f. Set the R/G Switch to R, and adjust the 1666 decade to the calibrated value of the 1440 Standard Resistor,
g. Readjust R107 for a zero meter deflection.
h. Set the R/G Switch to $G$ and read the value (reciprocal of R). It should be within 10 ppm . A final adjustment of R112 and possible re-adjustment of R107 may be necessary.

## Ratio Arm.

a. Set the 1666 front-panel controls as follows: BRIDGE SOURCE - Internal

WHEATSTONE/KELVIN - KELVIN
R/G-R
SENSITIVITY - as required
ZERO/MEASURE - MEASURE (perform zero adjust first)
METER RESPONSE - FAST
MULTBY $-100 \mathrm{~m} \Omega$
b. Connect a $1-\Omega$ GR Type 1440 Standard Resistor to the 1666 Bridge with the lead set for 4 -terminal measurements.
c. Set the decade to the calibrated value of the standard resistor, and adjust R101 to meter null.
d. Perform the YOKE adjustment (para 2.3.2), and repeat the R101 adjustment, if necessary.
e. Adjust the remaining bridge board potentiometers (range resistors) as listed in Table 5-3, using the procedure described in the last paragraph, except that the YOKE adjustment is not required above $1 \Omega$.

### 5.6 KNOB REMOVAL.

If it should be necessary to remove any knobs on the front-panel to replace one that has been damaged, or to replace the associated control, proceed as follows:
a. Grasp the knob firmly with the fingers close in to the panel (or MULT BY dial if applicable), and pull the knob straight away from the panel.

## CAUTION

Do not pull on the dial to remove a dial/knob assembly. Always remove the knob first. To avoid damage to the knob and other parts of the control, do not pry the knob loose with a screwdriver or similar flat tool, and do not attempt to twist the knob from the dial.
b. Observe the position of the setscrew in the bushing, with respect to any panel markings (or at the full ccw position of a continuous control).
c. Release the setscrew and pull the bushing off the shaft.

Table 5-3
RATIO ARM RESISTOR ADJUSTMENTS

| RATIO ARM RESISTOR ADIUSTMENTS |  |  |  |
| :---: | :---: | :---: | :---: |
| Standard | Mult By | R Adjust | Bridge Type |
| $1 \Omega$ | $100 \mathrm{~m} \Omega$ | $R 101$ | K |
| $10 \Omega$ | $1 \Omega$ | $R 103$ | K |
| $100 \Omega$ | $10 \Omega$ | $R 105$ | K |
| $1 \mathrm{k} \Omega$ | $100 \Omega$ | $R 107^{*}$ | W |
| $10 \mathrm{k} \Omega$ | $1 \mathrm{k} \Omega$ | R 109 | W |
| $100 \mathrm{k} \Omega$ | $10 \mathrm{k} \Omega$ | R 110 | W |
| $1 \mathrm{M} \Omega$ | $100 \mathrm{k} \Omega$ | R 111 | W |

* This adjustment was made previously in para 5.5 .7 , but can be readjusted here. However, any re-adjustment of R112 will require the procedure of all the ratio-arm adjustments to be repeated.

Table 5-4
NOMINAL TRANSISTOR VOLTAGES*

|  | Emitter | Base | Collector | Number | Junction Voltages |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Q1 | Source/OV | Gate/OV | Drain/OV | E101 | R6, R8/1.43 |
| Q2 | 0.9 | 1.4 | 5.1 | 2N4250 | R15, R18/2.1 |
| Q3 | 0.9 | 1.4 | 6.0 | 2N4250 | R23, R25/3.2 |
| Q4 | 6.0 | 5.1 | 2.8 | 2N3903 | R41, R42/4.0 |
| 05 | 1.2 | 1.7 | 5.4 | 2N3905 | R43, R59/4.2 |
| Q6 | 1.2 | 1.7 | 6.0 | 2N3905 |  |
| Q7 | 6.0 | 5.4 | 3.0 | 2N3903 |  |
| Q8 | 2.43 | 2.8 | 5.4 | 2N3905 |  |
| Q9 | 2.43 | 2.8 | 6.0 | 2N3905 |  |
| Q10 | 6.0 | 5.4 | 3.9 | 2N3903 |  |
| 011 | 3.2 | 3.9 | 5.8 | 2N3905 |  |
| 012 | 3.2 | 3.9 | 0 | 2N3903 |  |
| 013 | 0 | 0.35 | 3.4 | 2N3905 |  |
| 014 | Source/0 | Gate/3.4 | Drain/3.5 | E113 |  |
| 015 | 5.4 | 4.7 | 3.2 | 2N3903 |  |
| 016 | 5.4 | 4.7 | 3.2 | 2N3903 |  |

[^1]
## Parts Lists and Diagrams-Section 6

MECHANICAL PARTS LIST ..... 6-2
1666 BRIDGE COMPLETE ..... 6-3
DECADE SWITCH PARTS LIST ..... 6-3
DETECTOR BOARD PARTS LIST ..... 6-4
DETECTOR BOARD ETCHED AND CIRCUIT DIAGRAMS ..... 6-5
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## NOTE

Electrical parts information in this section is presented in such a way that all the data for a partnumbered subassembly is visible in a single opening of the manual. Thus the parts list appears on lefthand pages, while the etched circuit and schematic diagrams (tip out) are on the right-hand pages.


Figure 6-1. Replaceable mechanical Parts.

MECHANICAL PARTS LIST

| Fig Ref | Ont | Description | GR Part No. | Fed Mfy Code | Whaty Part No. | Fed Stoek No. |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1. | 1 | Panel gasket | 5331-3606 | 24655 | 5331-3606 |  |
| 2. | 1 | Switch, toggle AS5 ZERO MEASURE requires: | 7910-1810 | 39317 | 1111-0014 |  |
|  |  | dress nut | 5800-0800 | 24655 | 5800-0800 | 5310-344-3634 |
| 3. | 1 | Switch, toggle AS3 requires: | 7910-1820 | 39317 | 1111-0054 |  |
|  |  | dress nut | 5800-0800 | 24655 | 5800-0800 | 5310-344-3634 |
| 4. | 1 | Knob asm,, gray ZERO ADJ inc: | 5520-5220 | 24655 | 5520-5220 |  |
|  |  | retainer | 5220-5402 | 24655 | 5220-5402 |  |
|  |  | washer | 8140-0104 | 24655 | 8140-0104 |  |
| 5. | 2 | Knob asm., gray GUARD or YOKE, SENSITIVITY <br> inc: | 5520-5221 | 24655 | 5520-5221 |  |
|  |  | retainer | 5220-5402 | 24655 | 5220-5402 |  |
| 6. | 3 | Knob asm., gray METER RESPONSE POWER OFF, WHEATSTONE KELVIN inc: | 5500-5221 | 24655 | 5500-5221 |  |
|  |  | retainer | 5220-5402 | 24655 | 5220-5402 |  |
| 7. | 4 | Bushing, binding post | 0938-7130 | 24655 | 0938-7130 |  |
| 8. | 4 | Binding post, gold AJl-AJ4 | 4060-0108 | 24655 | 4060-0108 | 5905-912-0007 |
| 9. | 1 | Shorting link, gold | 0938-9503 | 24655 | 0938-9503 |  |
| 10. | 1 | Binding Post asmos AJ5 | 0938-2032 | 24655 | 0938-2032 |  |
| 11. | 1 | Spacer, binding post, gold | 0938-9830 | 24655 | 0938-9830 |  |
| 12. | 1 | Dial asm., outer Range Select requires: | 1666-1060 | 24655 | 1666-1060 |  |
|  |  | bushing | 4143-3253 | 24655 | 4143-3253 |  |
| 13. | 1 | Knob asm., gray Range Select inc: | 5520-5420 | 24655 | 5520-5420 |  |
|  |  | retainer | 5220-5401 | 24655 | 5220-5401 |  |
| 14. | 1. | Dial asm., inner $\mathrm{R}-\mathrm{G}$ Select | 1666-1110 | 24655 | 1666-1110 |  |
| 15. | 1 | Knob, gray R-G Select | 5540-3312 | 24655 | 5540-3312 |  |
| 16. | 10 | Knob, black Decade | 1650-5999 | 76854 | 294-29-001 |  |
| 17. | 1 | Cabinet asm. comp., inc: | 4182-2347 | 24655 | 4182-2347 |  |
|  | 1 | cabinet base complete | 4182-1347 | 24655 | 4182-1347 |  |
|  | 1 | handle asm. | 4182-1506 | 24655 | 4182-1506 |  |
|  | 1 | cover | 4182-8447 | 24655 | 4182-8447 |  |
|  | 1 | gasket | 5168-0680 | 24655 | 5168-0680 |  |
|  | 4 | foot | 5260-2060 | 24655 | 5260-2060 |  |
| 18. | 1 | Window, Meter | 5730-7000 | 24655 | 5730-7000 |  |
| MISCELLANEOUS |  |  |  |  |  |  |
| 3 |  | Bushing, Threaded AJ6-8 EXIT EXIT SOURCE, GROUND | 4150-2600 | 24655 | 4150-2600 | 6625-043-2108 |


| Ref Des | Fed |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | Description | GR Part No. | Mig Code | Mig Part No. | Fed Stock No. |
| CAPACITORS |  |  |  |  |  |
| ClOL | Electrolytic, $600 \mu \mathrm{~F},+150-10 \%, 3 \mathrm{~V}$ | 4450-5589 | 37942 | TCM, $600 \mu \mathrm{~F}, 3 \mathrm{~V}$ | 5910-929-9975 |
| METERS |  |  |  |  |  |
| AM1 | Meter | 5730-1666 | 24655 | 5730-1666 |  |
| RESISTORS |  |  |  |  |  |
| ARI | Potentiometer, Comp., $250 \Omega, \pm 10 \%$ | 6000-0100 | 01121 | $\mathrm{JU}, 250 \Omega, \pm 10 \%$ |  |
| AR2 | Potentiometer, Comp., $10 \mathrm{~T}, 1 \mathrm{M} \Omega \pm 20 \%$ | 6045-0460 | 01121 | JJ, $10 \mathrm{~T}, 1 \mathrm{M} \Omega, \pm 20 \%$ |  |
| AR3 | Potentiometer, Comp., $10 \mathrm{k} \Omega \pm 20 \%$ | 6020-0400 | 01121 | $J \mathrm{~J}_{2}, 10 \mathrm{k} \Omega, \pm 10 \%$ | 5905-829-3323 |
| AR151 | Film, $110 \mathrm{k} \Omega, \pm 1 \%, 1 / 8 \mathrm{~W}$ | 6250-3110 | 75042 | CEA, $110 \mathrm{k} \Omega, \pm 1$ \% |  |
| AR152 | Composition, $200 \Omega, \pm 5 \%, 1 / 4 \mathrm{~W}$ | 6099-1205 | 75042 | BTS, $200 \Omega, \pm 5 \%$ | 5905-683-2239 |
| AR153 | Composition, $62 \Omega, \pm 5 \%, 1 / 4 \mathrm{~W}$ | 6099-0625 | 75042 | BTS, $62 \Omega, \pm 5 \%$ |  |
| AR154 | Composition, $1 \mathrm{k} \Omega, \pm 5 \%, 1 / 4 \mathrm{~W}$ | 6099-2105 | 75042 | BTS, $1 \mathrm{kn}, \pm 5 \%$ | 5905-681-6422 |
| AR155 | Composition, $10 \Omega, \pm 5 \%, 1 / 4 \mathrm{~W}$ | 6099-0105 | 75042 | BTS, $10 \Omega, \pm 5 \%$ | 5905-809-8596 |
| AR156 | Composition, $100 \mathrm{k} \Omega, \pm 5 \%, 1 / 4 \mathrm{~W}$ | 6099-4105 | 75042 | BTS, $100 \mathrm{k} \Omega \pm 5 \%$ | 5905-686-3129 |
| AR157 | Wire Wound, $1.0 \Omega, 45 \%, 2 \mathrm{~W}$ | 6760-9105 | 75042 | BWH, $1.0 \Omega, \pm 5 \%$ |  |
| AR158 | Composition, $750 \Omega, \pm 5 \%, 1 / 4 \mathrm{~W}$ | 6099-1755 | 75042 | BTS, $750 \Omega, \pm 5 \%$ |  |
| AR160 | Composition, $470 \mathrm{k} \Omega, \pm 5 \%, 1 / 4 \mathrm{~W}$ | 6099-4475 | 75042 | BTS, $470 \Omega, \pm 5 \%$ |  |
| AR161 | Composition, $10 \mathrm{k} \Omega, \pm 5 \%, 1 / 4 \mathrm{~W}$ | 6099-3105 | 75042 | BTS, $10 \mathrm{k} \Omega, \pm 5 \%$ | 5905-683-2238 |
| SWITCHES |  |  |  |  |  |
| AS1 | Rotary, Wafer | 7890-5605 | 76854 | 7890-5605 |  |
| AS3 | Toggle | 7910-1810 | 39317 | 1111-0014 |  |
| AS5 | Toggle | 7910-1820 | 39317 | 1111-0054 |  |
| AS6 | Rotary, Wafer | 7890-5604 | 76854 | 7890-5604 |  |
| AS7 | Rotary, Wafer | 7890-5606 | 76854 | 7890-5606 |  |

DECADE SWITCH ASM. COMP.

SWITCH ASM: $50 \mathrm{~m} \Omega / \mathrm{step}$

## RESISTORS

R1 thru
R6 Pwr., Wire Wound, $0.1 \Omega, \pm 1 / 2 \%, 1 \mathrm{~W}$ 6620-1031 75042 WW3J, $1 \Omega, \pm 1 / 2 \%$
SWITCH ASM: $.5 \Omega /$ step
RESISTORS
RI thru
R6 Film, $1 \Omega, \pm 1 \%, 300 \mathrm{ppm} \quad 6619-2803$

R12 Pwr., Wire Wound, $0.1 \Omega, \pm 5 \%, 1 \mathrm{~W}$ 6620-1032
75042
$\mathrm{CEA}-\mathrm{TO}, 1 \Omega, \pm 1 \%$

SWITCH ASM: $5 \Omega /$ step
RESISTORS
RI thru
R6 Film, $10 \Omega, \pm 0.1 \%, 501 / 8 \mathrm{~W}$
R12 Film, $2 \Omega_{2} \pm 1 \%, 300 \mathrm{ppm}$
6190-0001 750
R 13 and Film, $2 \Omega, \pm 1 \%, 300 \mathrm{ppm}$
R14 Comp, $20 \Omega, \pm 5 \%, 1 / 4 \mathrm{~W}$
6619-2805 7504
MEA, $10 \Omega, \pm 0.1 \%$

6099-0205
75042
BTS, $20 \Omega, \pm 5 \%$

SWITCH ASM: $50 \Omega /$ step
RESISTORS
R1 thri

| R6 | Film, $100 \Omega, \pm 0.1 \%, 100 \mathrm{ppm}$ | $6619-1600$ | 75042 | CEA-TO, $100 \Omega, \pm 0.1 \%$ |
| :--- | :--- | ---: | ---: | ---: |
| R7 thru |  |  |  |  |
| R12 | Film, $20 \Omega, \pm 0.1 \%, 501 / 8 \mathrm{~W}$ | $6190-0004$ | 75042 | MEA, $20 \Omega, \pm 0.1 \%$ |

SWITCH ASM: $500 \Omega /$ step
RESISTORS
R1 thru
R6 Wire Wound, 1 K, 1K, $1 \mathrm{~K} \quad 6990-3210 \quad 24655 \quad 6990-3210$
SWITCH ASM: $5 \mathrm{~K} /$ step
RESISTORS

| R1 thru |  |  |  |  |
| :--- | :--- | :--- | :--- | :--- |
| R6 | Wire Wound, $10 \mathrm{~K}, 10 \mathrm{~K}, 10 \mathrm{~K}$ | $6990-3310$ | 24655 | $6990-3310$ |
| R7 thru |  |  |  |  |
| R12 | Film, $2 \mathrm{k} \Omega_{8} \pm .1 \%$ | $6619-1120$ | 75042 | CEA-TO, $2 \mathrm{k} \Omega, \pm 1 \%$ |

```
*)
```




| capactimis |  |
| :---: | :---: |
| Cl | Mylar, $1 \mu \mathrm{~F}, \pm 10 \%, 200 \mathrm{~V}$ |
| $\mathrm{C}_{3}$ and |  |
|  |  |
| $\begin{gathered} \mathrm{C4} \\ \hline \end{gathered}$ | Mylar, $.01 \mu \mathrm{~F}, \pm 2 \%$, 100 |
| C5 Ceramic, $01 \mu \mathrm{~F},+80-20 \%$, 100 V |  |
|  | Mylar, . $01 \mu \mathrm{~F}, \pm 2 \%$, |
| C8 Ceramic, $01 \mu \mathrm{~F},+80-20 \%$, 100 V |  |
| $\mathrm{ClO}^{1}$ Mylar, $.01 \mu \mathrm{~F}, \pm 2 \%$, 100 V |  |
| Cll and | Mylar, .01 $\mu \mathrm{L}$ |
| $\mathrm{Cl2}^{\text {Ele }}$ |  |
| C14 and |  |
| ${ }_{C} 15$ | Electrolytic, $200 \%$, $+150-10 \% 6 \mathrm{~V}$ |
| Tantalum, 6.8 |  |
|  |  |
|  |  |
|  |  |
| C20 and |  |
| C21 | Tantalum, $6.8 \mu \mathrm{~F}, \pm 20 \%$, |
| C22 | Ceramic, $01 \mu$ |
| ${ }^{2} 23$ Ceramic, $1000 \mathrm{pF}, 110 \%$ |  |
| C24 and |  |
|  |  |
| ${ }^{\text {c26 }}$ - ${ }^{\text {c }}$ |  |
| ${ }_{\mathrm{C} 28}{ }^{\text {and }}$ Electrolytic, $50 \mu \mathrm{~F},+150-10 \%$, |  |
|  |  |
| C29 and |  |
|  | Tantalum, $6.8 \mu \mathrm{~F}, \pm 20 \%, 6 \mathrm{~V}$ |
| diodes |  |
| $\mathrm{CR1}^{\text {and }}$ |  |
|  | Type 1N4009 |
| ${ }_{\text {CR6 }}$ CR ${ }^{\text {and }}$ Type IN118A |  |
|  |  |
| CR7 Type 1N118A |  |
| CR8 thru |  |
| CR13 | Type in4009 |
| RESISTORS |  |
|  | Film, $10 \mathrm{k} \Omega, \pm 1 \%, 1 / 8$ |
| R2 | Composition, $3 \mathrm{M} \Omega$, |
| R3 | Composition, $1 \mathrm{M} \Omega, 55 \%, 1 / 4 \mathrm{~W}$ |
| R4 | Comp., 47 k R, $\pm 5 \%$, $1 / 4 \mathrm{~W}$ |
| R5 | Composition, $22 \mathrm{k} \mathrm{\Omega}, \pm 5 \%, 1$ |
| R6 | Film, $348 \mathrm{kK}, \pm 1 \mathrm{l}, 1 / 8$ |
| R7 | Film, $1.27 \mathrm{k},{ }^{\text {, }} \pm 1 \%$ \% $1 / 8 \mathrm{~W}$ |
| R8 | Composition, $10 \mathrm{k} \Omega, 55 \%, 1 / 4 \mathrm{~W}$ |
| R9 | Composition, $2 \mathrm{k} \Omega_{\text {, }} \pm 5 \%, 1 / 4 \mathrm{~W}$ |
| R11 | Composition, $1 \mathrm{M} \Omega$ |
| $\mathrm{Rl}^{12}$ | Composition, $470 \mathrm{kS}, \pm 55,1 / 4 \mathrm{~W}$ |
| R13 | Composirion, $22 \mathrm{k} \Omega_{\text {, }}+5 \%, 1 / 4 \mathrm{~W}$ |
| R14 | Composition, $10 \mathrm{k} \Omega^{2}, 55 \%, 1 / 4 \mathrm{~W}$ |
| R15 | Film, $348 \mathrm{kS}, \pm 1 \%$, $1 / 8$ |
| R16 | Film, $1.27 \mathrm{k} \Omega, \pm 1 \%$, $1 / 8 \mathrm{~W}$ |
| R17 | Composition, $4.7 \mathrm{kR}, \pm 5 \%, 1 / 4 \mathrm{~W}$ |
| R19 and |  |
|  |  |
| ${ }_{\mathrm{R} 21}^{\mathrm{R} 20}$ and Composition, $1 \mathrm{M} \Omega$, |  |
|  |  |
| $\stackrel{\text { R23 }}{\text { R23 }}$ | Composition, $22 \mathrm{k} \Omega, \pm 5 \%, 1 / 4 \mathrm{~W}$ |
| ${ }_{\text {R }}$ | Film, $348 \mathrm{ka}, \pm 1$, |
| R25 | Composition 22 kg |
| 26 | Composition 2 k |
| 7 | Compostion, 2100 |
| 8 | on, |
| R28 | Composition, $100 \mathrm{k}, 2,25 \%, 1 / 4$ |
| ${ }^{\text {thru }}$ |  |

$\begin{array}{lll}4860-8274 & 84411 & 663 \mathrm{UW}, 1 \mu \mathrm{~F} \pm 10 \% \\ 4860-8200 & 84411 & 663 \mathrm{UW}, 047 \mu \mathrm{~F}, \pm 10 \%\end{array}$
$\begin{array}{lllll}4860-7650 & 84411 & 663 \mathrm{UW}, .01 \mu \mathrm{~F}, \pm 2 \% & \\ 4401-3100 & 80131 & \mathrm{CC61,0}, 01 \mu \mathrm{~F},+80-20 \% & 5910-974-5697\end{array}$

$4860-7650 \quad 84411 \quad 663 \mathrm{UW}, .01 \mu \mathrm{~F}, \pm 2 \%$
$\begin{array}{llll}{ }^{4450-5616} & 56289 & \text { 150D127XXOMOR2 } \\ 4450-4800 & 56289 & 150 D 655 X 0010 \mathrm{~A} 2 & 5910-936-1332\end{array}$
 $\begin{array}{llll}4450-4800 & 56289 & 150 D 685 \times 0010 \mathrm{~A} 2 & \\ 8960-7853 & 84411 & 5930-936-1332\end{array}$
$\begin{array}{lllll}4450-4800 & 56289 & 150 \mathrm{D} 685 \times 0010 \mathrm{~A} 2 & & 5910-936-1332 \\ 4401-3100 & 80131 & \mathrm{CCC1}, 01\end{array}$
$\begin{array}{lllll}4401-3100 & 80131 & \mathrm{CC} 61, & 01 \mu \mathrm{~F},+80-20 \% & 5910-974-5697\end{array}$
$410-0475 \quad 80131 \quad \mathrm{CC} 61,47 \mathrm{pF}, \pm 5$
$4450-5590 \quad 37942 \quad$ TT, $50 \mu \mathrm{~F}, 3 \mathrm{~V}$
$4450-4800 \quad 56289 \quad 1507685 \times 0010 \mathrm{~A} 2 \quad 5910-936-1332$

| 6082-1012 | 24446 | ${ }^{1 N 4442}$ | 5961-929-9967 |
| :---: | :---: | :---: | :---: |
| 6082-1006 | 98925 | 1N118A |  |
| 6082-1006 | 98925 | 1N118A |  |
| 6082-1012 | 24446 | 1N4442 | 5961-929-9967 |
| 6250-2100 | 75042 | CEA, $10 \mathrm{k} \Omega$, | 5905-883 |
| 6099-5305 | 75042 | BTS, 3 M , |  |
| $6099-5105$ |  | STS, $1 \mathrm{M} \Omega, \pm 5 \%$ |  |
| 6099-3475 <br> 6099-322 | ${ }_{75042}^{75042}$ | ${ }_{\text {BTS }} \mathrm{BTS}, 27 \mathrm{k} \mathrm{kR}^{2}+55 \%$ | 5905-687-0002 |
| . $6250-3348$ | 75042 | CEA-TO, $348 \mathrm{k} 2, \pm 1 \%$ |  |
| $6250-1127$ | 75042 | CEA, $1.27 \mathrm{k} \Omega, \pm 1 \%$ | 5905-771-0281 |
| 6099-31 | 75042 | BTS, $10 \mathrm{kR}, \pm 5 \%$ | 5905-686-3370 |
| ${ }^{6099-2205}$ | 75042 | BTS, $2 \mathrm{kR}, \pm 5 \%$ |  |
| 6099-5105 | ${ }_{75042}^{75042}$ |  |  |
| 6099-3225 | 75042 | BTS, $22 \mathrm{k} \Omega_{2} \pm 5 \%$ | 5905-687-0002 |
| 6099-3105 | 75042 | BTS, $10 \mathrm{k} \Omega, \pm 5 \%$ | 5905-683-2238 |
| 6250-3348 | 75042 | CEA-TO, $348 \mathrm{k} \Omega, \pm 1 \%$ |  |
| 6250-1127 | 75042 | CEA, $1,27 \mathrm{kS}, \pm 1{ }^{\text {m }}$ | 5905-721-0281 |
| 609 | 75042 | BTS, $4.7 \mathrm{k} \Omega, 55 \%$ | 5995-686-9992 |
| 6099-3105 | 75042 | BTS, $10 \mathrm{k} \Omega, \pm 5 \%$ | 5905-683-2238 |
| 6099-5105 | 7504 | BTS, 1 M $8, \pm 5 \%$ |  |
| 6099-3225 | 75042 | BTS, $22 \mathrm{k} \Omega$, | 5905-687-0002 |
| ${ }^{6250-3348} \mathbf{6 2 0 0 - 1 1 2 7}$ | 75042 | ${ }_{\text {CEA }} \mathrm{CEA}, 1.27 \mathrm{k}, 38 \mathrm{k}, \pm 1 \%$ | 5905-721-0281 |
| 6099-3225 | 75042 | BTS, 22 k , $\pm 5 \%$ | 5905-687-0002 |
| 6099-2205 | 75042 | BTS, $2 \mathrm{k} \Omega, \pm 5 \%$ | 5905-686-3370 |
| ${ }^{6099-1105}$ | 75042 | BTS, $100 \Omega, \pm 5 \%$ | 5905-686-3129 |
| 6099-4105 |  |  |  |
| 6099-2205 | 75042 | BTS, $2 \mathrm{k} \Omega$, $\pm 5 \%$ | 5905-686-3370 |




NOTE
The board is shown foil-side up. The number appearing on the foil side is not the part number. Symbolism Gray area = part; black circuit pattern = etch; Square pad in ckt pattern = collector, cathode (of diode), or + end of capcaitor.

| Ref Des | Description | GR Part No. | Fed Mfg Code | Mfg Part No. | Fed Stock No. |
| :---: | :---: | :---: | :---: | :---: | :---: |
| R121 | Wire Wound, $1.0032 \Omega, \pm .025 \%, 1 \mathrm{~W}$ | 6983-1005 | 24655 | 6983-1005 |  |
| R122 | Wire Wound, $10.32 \Omega, \pm .025 \%, 1 \mathrm{~W}$ | 6983-2003 | 24655 | 6983-2003 |  |
| R123 | Wire Wound, $100.32 \Omega, \pm .025 \%, 1 \mathrm{~W}$ | 6983-3015 | 24655 | 6983-3015 |  |
| R124 | Wire Wound, $1.010 \mathrm{k} \Omega, \pm .025 \%, 1 \mathrm{~W}$ | 6983-4042 | 24655 | 6983-4042 |  |
| R125 | Wire Wound, $9.995 \mathrm{k} \Omega, \pm .025 \%$, 1 W | 6983-4043 | 24655 | 6983-4043 |  |
| R126 | Wire Wound, $99.95 \mathrm{k} \Omega, \pm .025 \%$, 1 W | 6991-2902 | 24655 | 6991-2902 |  |
| R127 | Wire Wound, $999.5 \mathrm{k} \Omega, \pm .025 \%, 1 \mathrm{~W}$ | 6991-2903 | 24655 | 6991-2903 |  |
| R131 | Composition, $180 \Omega, \pm 5 \%, 1 / 4 \mathrm{~W}$ | 6099-1185 | 75042 | BTS, $180 \Omega, \pm 5 \%$ | 5905-682-4107 |
| R132 | Composition, $18 \mathrm{k} \Omega, \pm 5 \%, 1 / 4 \mathrm{~W}$ | 6099-3185 | 75042 | BTS, $18 \mathrm{k} \Omega$, $\pm 5 \%$ | 5905-687-0000 |
| R133 | Composition, $1.8 \mathrm{M} \Omega, \pm 5 \%, 1 / 4 \mathrm{~W}$ | 6099-5185 | 75042 | BTS, $1.8 \mathrm{M} \Omega, \pm 5 \%$ | 5905-688-3738 |
| R134A and a |  |  |  |  |  |
| R134B | Composition, $100 \mathrm{M} \Omega, \pm 10 \%, 1 / 2 \mathrm{~W}$ | 6100-7105 | 01121 | EB, $120 \Omega, \pm 5 \%$ |  |
| R141 | Composition, $120 \Omega, \pm 5 \%, 1 \mathrm{~W}$ | 6110-1125 | 01121 | RC32GF121J | 5905-279-1726 |
| R142 | Composition, $430 \Omega, \pm 5 \%, 1 \mathrm{~W}$ | 6110-1435 | 01121 | RC32GF431J |  |
| R143 | Composition, $820 \Omega, \pm 5 \%, 1 / 2 \mathrm{~W}$ | 6100-1825 | 01121 | RC32GF821J | 5905-279-2651 |
| R144 thru |  |  |  |  |  |
| R146 | Composition, $1 \mathrm{k} \Omega, \pm 5 \%, 1 \mathrm{~W}$ | 6110-2105 | 01121 | RC32GF102J | 5905-473-5251 |
| R162 | Film, $10 \mathrm{k} \Omega, \pm 0.1 \%, \pm 50 \mathrm{ppm} 1 / 2 \mathrm{~W}$ | 6188-2100 | 75042 | MEC-T2, $10 \mathrm{k} \Omega, \pm 1 \%$ |  |
| R163 | Film, $453 \mathrm{k}, \pm 1 \%, 1 / 4 \mathrm{~W}$ | 6350-3453 | 75042 | CEB, $453 \mathrm{k} \pm 1 \%$ |  |



Figure 6-5. Standard arm switch assembly (AS4, AS8) (P/N 1666-2050).



| A-S4,405 FER |  | AS9,305F |  | BY R141 |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| A-S4,407FER |  | AS4,307F |  | BY | R142 |  |  |
| ASS 4, 409 FE R | TO | ASA,309F |  | BY | R143 |  |  |
| A-S4,411FER |  | $A-54,311 F$ |  | BY | R194 |  |  |
| A-S4, 413 FER | TO | AS4,313F |  | BY | $18145^{\circ}$ |  |  |
| AS $4,415 \mathrm{FER}$ | TO | AS4,315F |  | BY | R/46 |  |  |
| A-54,203R | TO | A-S4,303F |  | By | R121 |  |  |
| AS4,203R | TO | $A-54,304 R$ |  | BY | R/31 |  |  |
| AS4,205R | TO | AS4,305F |  | BY | R122 |  |  |
| A-S4,205R | TO | A-S4,306R |  | BY | R/32 |  |  |
| A-S4,207R |  | A-54,307F |  | BY | R123 |  |  |
| A-S4,207R | TO | AS4,308R |  | $B Y$ | R133 |  |  |
| A.54,209R | TO | AS4,309F |  | BY | R124 |  |  |
| 4.54,209R |  | A-54,310R |  |  | R134A ' R 3 (38 |  |  |
| A-S4,212R |  | A-54,311F |  | By | R125 |  |  |
| A-S4,214R | 70 | 4.54,313F |  | $B Y$ | R126 |  |  |
| A-S4,2/6R | TO | 4-54,315F |  | $B Y$ | R127 |  |  |
| A-SE, $105 R$ | TO A-S8,107 |  |  |  | R/62,R/63 |  |  |
| A-SB,1/6R |  | A-S8,103FER |  | $B Y$ | R/61 |  |  |
| A-S4,101F |  | A-S4,102F |  |  |  |  |  |
| A-S4,101R |  | A-S4,102R |  |  |  |  |  |
| A-S4,201F | TO | A-S4, 202F |  |  |  |  |  |
| A-54,201R | TO A-s4,202R |  |  |  |  |  |  |
| A-54,103F TO |  | S4,103R | TO | A-S4, | 2037 | TO | AS4,203R |
| A-S4,105\% T0 |  | S4,105R | TO | A-54.20 | 2055 | 70 | A54,205R |
| A-54,107F TO |  | S4,107R | TO | A-54, | 207\% | T0 | A-S4,207R |
| A-S4,109F TO | A- | S4,109R | TO | A-54 | 209F | T0 | A-S4,209R |
| A-SQ,IIIF TO | A-S | Sq, I11R | TO | A-S4, | $211 F$ | TO | A-S $4,211 R$ |
| A-S4,13F TO | A-S | S4,113R | TO | A-S4, | 2135 | T0 | A-S4,213R |
| A-S4,115 | A- | S4,15R |  | A-54 | $215{ }^{\circ}$ | TO | A-S4,215K |
| A-54,303F TO |  | S4,403F\% |  |  |  |  |  |
| A-S8,102F TO |  | S8, 102R |  |  |  |  |  |
| A-S8,103F TO |  | S8,103R |  |  |  |  |  |
| A-58,107F TO |  | 58,107R |  |  |  |  |  |
| A-58,108F 70 |  | S8,108R |  |  |  |  |  |
| A-58,113F TO |  | S8,113R |  |  |  |  |  |
| A-S8,114F TO |  | S8,114R |  |  |  |  |  |
| $A-S 8,101 R$ TO | A-S | 54,302F |  |  |  |  |  |

FEDERAL MANUFACTURER'S CODE
From Federal Supply Code for Manufacturers Cataloging Handbooks H4-1 (Name to Code) and H4-2 (Code to Name) as supplemented through August, 1968.

| Code | Manufacturer |
| :---: | :---: |
| 00192 | Jones Mfg. Co, Chicago, Illinois |
| 00194 | Walsco Electronics Corp, L.A., Calif. |
| 00434 | Schweber Electronics, Westburg, L.,., N.Y. |
| 00656 | Aerovox Corp, New Bedford, Mass. |
| 00779 | Amp Inc., Harrisburg Pa., 17105 |
| 01009 | Alden Products Co, Brockton, Mass. |
| 01121 | Allen-Bradley, Co, Milwaukee, Wisc. |
| 01236 | Leeds Radio Company, N.Y |
| 01255 | Litton Industries Inc, Beverly Hills, Calif. |
| 01295 | Texas Instruments, Inc, Dallas, Texas |
| 02114 | Ferroxcube Corp, Saugerties, N.Y. 12477 |
| 02606 | Fenwal Lab inc, Morton Grove, ill. |
| 02660 | Amphenol Electron Corp, Broadview, III. |
| 02768 | Fastex, Des Plaines, Ill. 60016 |
| 03042 | Carter Ink Co., Camb. Mass. 02142 |
| 03508 | G.E. Semicon Prod, Syracuse, N.Y. 13201 |
| 03636 | Grayburne, Yonkers, N.Y. 10701 |
| 03888 | Pyrotilm Resistor Co, Cedar Knotls, N.J. |
| 03911 | Clairex Corp, New York, N.Y. 10001 |
| 04009 | Arrow-Hart \& Hegeman, Hart., Conn. 06106 |
| 04643 | Digitronics Corp., Albertson, N.Y. 11507 |
| 04713 | Motorola, Phoenix, Ariz. 85008 |
| 05170 | Engr'd Electronics, Santa Ana, Calif. 92702 |
| 05624 | Barber-Colman Co, Rockford, III. 61101 |
| 05748 | Barnes Mfg. Co., Mansfield, O. 44901 |
| 05820 | Wakefield Eng, Inc, Wakefield, Mass. 01880 |
| 06743 | Clevite Corp., Cleveiand, O. 44110 |
| 06751 | Nuclear Corp., of America, Inc., Phoenix, Ariz |
| 07126 | Digitron Co, Pasadena, Calif. |
| 07127 | Eagle Signal (E.W. Bliss Co.), Baraboo, Wisc. |
| 07233 | Cinch-Graphik, City of tidustry, Calif. |
| 07261 | Avnet Corp, Culver City, Calif. 90230 |
| 07263 | Fairchild Camera, Mountain View, Calif. |
| 07387 | Birtcher Corp, No. Los Angeles, Calif. |
| 07595 | Amer Semicond, Arlington Hts, 111. 60004 |
| 07828 | Bodine Corp, Bridgeport, Conn. 06605 |
| 07829 | Bodine Electric Co, Chicago, III. 60618 |
| 07910 | Cont Device Corp, Hawthorne, Calif. |
| 07983 | State Labs Inc, N.Y., N.Y. 10003 |
| 07999 | Borg tnst., Delavan, Wisc. 53115 |
| 08730 | Vemaline Prod Co., Franklin Lakes, N.J. |
| 09213 | G.E. Semiconductor, Buffalo, N.Y. |
| 09408 | Star-Tronics Inc, Georgetown, Mass 01830 |
| 09823 | Burgess Battery Co, Freeport, III. |
| 09922 | Burndy Corp, Norwalk, Conn. 06852 |
| 11236 | C.T.S. of Berne, Inc, Berne, Ind. 46711 |
| 11599 | Chandler Evans Corp, W. Hartford, Conn. |
| 12040 | National Semiconductor, Danbury, Conn. 120 |
| 12065 | Transitron Electronic Corp., E. Boston, Mass. |
| 12498 | Crystaionics, Cambridge, Mass. 02140 |
| 12617 | Homlin, Inc., Lake Millis, Wisc. 53551 |
| 12672 | RCA, Woodibridge, N.J. |
| 12697 | Clarostat Mfg Co, Inc, Dover, N.H. 03820 |
| 12954 | Dickson Electronics, Scottsdale, Ariz. |
| 13327 | Solitron Devices, Tappan, N.Y. 10983 |
| 14433 | ITT Semiconductors, W. Palm Beach, Fla. |
| 14655 | Cornell-Dubilier Electric Co., Newark, N.J. |
| 14674 | Corning Glass Works, Corning, N.Y. |
| 14936 | General Instrument Corp, Hicksville, N.Y. |
| 15116 | Microdot Magnetics Inc, Los Angeles, Calif. |
| 15238 | ITT, Semiconductor Div, Lawrence, Mass. |
| 15605 | Cutler-Hammer Inc, Milwaukee, Wisc, 53233 |
| 16037 | Spruce Pine Mica Co. Spruce Pine, N.C. |
| 16636 | Indiana General Corp, Oglesby, III. 61348 |
| 17771 | Singer Co, Diehl Div, Somerville, N.J. |
| 17856 | Siliconix, Inc., Sunnyvale, Calif. 94086 |
| 18736 | Voltronics Corp, Hanover, N.J. 07936 |
| 19396 | Ullinois Tool Works, Pakton Div, Chicago, Ill. |
| 19048 | Computer Diode Corp, S. Fairlawn, N.J. 07410 |
| 19617 | Cabtron Corp., Chicago, III. 60622 |
| 19644 | LRC Electronics, Horseheads, N.Y. |
| 19701 | Electra Mfg Co, Independence, Kansas 67301 |
| 20754 | KMC Semiconductor Corp., Long Valley, N.J. 07853 |
| 21335 | Fafnir Bearing Co, New Briton, Conn. |
| 22753 | UID Electronics Corp, Hollywood, Fla. |
| 23342 | Avnet Electronics Corp, Franklin Park, IIt. |
| 24446 | G.E., Schenectady, N.Y. 12305 |
| 24454 | G.E., Electronics Comp, Syracuse, N.Y. |
| 24455 | G.E. (Lamp Div.), Nela Park, Cleveland, Ohio |
| 24655 | General Radio Co, W. Concord, Mass. 01781 |
| 26806 | American Zettlet Inc, Costa Mesa, Calif. |
| 28520 | Hayman Mfg Co, Kenilworth, N.J. |
| 28959 | Hoffman Electronics Corp, El Monte, Calif. |
| 30646 | Beckman Instruments inc, Cedar Grove, N.J. 07009 |
| 30874 | 1.B.M., Armonk, New York |
| 32001 | Jensen Mfg. Co, Chicago. III. 60638 |
| 33173 | G.E. Comp, Owensboro, Ky 42301 |
| 34141 | Koehler Mfg. Co. Inc., Mariboro, Mass. 01752 |
| 35929 | Constanta Co, Mont. 19, Que. |

35929

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[^0]:    * $W=$ Wheatstone, $K=$ Kelvin, $R=$ Resistance, $G=$ Conductance.
    ** Calibrated standards should be used and their corrections applied.

[^1]:    *All measurements are in negative dc, volts with respect to ground.

